

PATENT APPLICATION

REALTIME ADAPTIVE NOTCH COMPENSATOR

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CROSS-REFERENCE TO RELATED APPLICATIONS

Not Applicable.

STATEMENT REGARDING FEDERALLY SPONSORED RESEARCH OR DEVELOPMENT

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Not Applicable.

INCORPORATION BY REFERENCE OF MATERIAL SUBMITTED ON A COMPACT DISC

Not Applicable.

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COPYRIGHTED MATERIAL

Not Applicable.

BACKGROUND OF THE INVENTION

Field of the Invention (Technical Field):

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The present invention relates to methods and apparatuses for reducing effects of 1) structural resonances in the control of mechanical structures, such as gimbaled turrets and 2) noise harmonics of time varying and/or uncertain frequency, such as the spin frequency noise in rate gyroscopes subject to power fluctuations.

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Description of Related Art:

In the control of mechanical structures, such as gimbaled turrets, the frequency of structural resonances must be known so they can be cancelled with notch filters in the controller to keep the

feedback control loop stable. The present invention adaptively computes the frequency of the resonance in realtime during normal closed-loop operation of the controller to improve the accuracy of the frequency estimate and adapt to changes with time in the resonance frequency. With a rough initial estimate, the Realtime Adaptive Notch Compensator (RANC) of the invention solves for a more
5 accurate estimate of the frequency and compensates the notch filter to stabilize the control loop thus minimizing oscillations due to the structural resonance.

Structural resonances are problematic in that they are very narrow in bandwidth, they vary from system to system even with identical structure design, and, in a given system, may also vary with gimbal
10 position, temperature, and vehicle high-g maneuvers.

The conventional approach is to measure and characterize each system to determine the frequency at which structural resonances occur. Constant, dedicated notches are then placed at these frequencies to attenuate the resonance amplitude. These notches must be narrow in width; otherwise
15 they will significantly degrade the controller's performance, in particular, the phase and gain stability margins. The emphasis on narrow notches increases the accuracy with which the structural frequency must be known, thus, making the measurement process lengthy and expensive.

The frequency measurement process can be extended to generate look-up tables of frequency
20 versus gimbal position, temperature, vehicle acceleration, etc. These tables can then be implemented in an open-loop fashion in the controller to set the notch frequency depending on gimbal position, temperature, vehicle acceleration, etc. The open-loop nature, however, provides no feedback indication that the frequency was adjusted properly to minimize oscillations.

25 Another approach is to minimize the interaction of structural resonances with the controller by designing stiffer structures such that resonances occur at higher frequencies where they cannot be excited by the controller. The disadvantage is larger, heavier structures or structures built with

expensive materials. Alternately, the controller bandwidth is kept low (at the expense of performance) such that the controller does not excite structural resonances.

Notch filters are also used to filter out noise harmonics in sensors such as spin frequency noise
5 in rate gyroscopes. The RANC of the invention is also applicable in such application to compensate for variations in spin frequency due, for example, to temperature changes or power fluctuations.

The prior art approach is exemplified schematically in Fig. 1, wherein an input signal **10** to be measured is corrupted by sensor **12** with harmonic noise of constant frequency ω_n , which is processed
10 by notch filter **14** at fixed frequency ω_n , resulting in filtered output **16** and notch output error **18**. The notch output error **18** is calculated via summing node **19** as the difference between the true input signal **10** and the filtered output **16**.

BRIEF SUMMARY OF THE INVENTION

15 The present invention is of a notch compensator comprising: means for receiving input to a notch filter; means for receiving output from the notch filter; means for dynamically calculating a desired change to a notch frequency of the notch filter; and means for specifying the desired change to the notch filter. In the preferred embodiment, the input receiving means comprises a first pre-filter providing a notch output error as output, the output receiving means comprises the first pre-filter, and the input
20 receiving means comprises a second pre-filter providing a reference signal as output. The calculating means comprises demodulation means receiving input from the first and second pre-filters, with preferably the demodulation means providing a frequency error as output. The calculating means additionally comprises integral compensation means receiving the frequency error as input, preferably wherein the integral compensation means minimizes the notch output error. The demodulation means
25 most preferably comprises a low pass filter that is second order with a bandwidth about one decade below an expected value of a next notch frequency.

The present invention is also of a notch compensation method comprising: receiving input to a notch filter; receiving output from the notch filter; dynamically calculating a desired change to a notch frequency of the notch filter; and specifying the desired change to the notch filter. In the preferred embodiment, receiving input comprises employing a first pre-filter that provides a notch output error as
5 output, receiving output comprises employing the first pre-filter, and receiving input comprises employing a second pre-filter that provides a reference signal as output. Calculating comprises demodulating employing input from the first and second pre-filters, preferably wherein the demodulating step provides a frequency error as output. Calculating preferably additionally comprises performing integral compensation employing the frequency error as input, most preferably wherein performing
10 integral compensation minimizes the notch output error. Demodulating most preferably comprises employing a low pass filter that is second order with a bandwidth about one decade below an expected value of a next notch frequency.

The present invention is additionally of a notch filtering system comprising a notch filter and a
15 notch compensator as described above. The system is useful in reducing structural resonances in control of mechanical structures, reducing noise harmonics of time varying and/or uncertain frequency, adaptive harmonic noise identification, adaptive harmonic noise filtering, and control of flexible structures. The system is also useful in control of gimbaled turrets, control of helicopters, stabilization platforms, gyroscopic rate sensors, computer hard drives, vehicle body bending compensation, and
20 flexible robotic manipulators.

Objects, advantages and novel features, and further scope of applicability of the present invention will be set forth in part in the detailed description to follow, taken in conjunction with the accompanying drawings, and in part will become apparent to those skilled in the art upon examination
25 of the following, or may be learned by practice of the invention. The objects and advantages of the invention may be realized and attained by means of the instrumentalities and combinations particularly pointed out in the appended claims.

BRIEF DESCRIPTION OF THE SEVERAL VIEWS OF THE DRAWINGS

The accompanying drawings, which are incorporated into and form a part of the specification, illustrate one or more embodiments of the present invention and, together with the description, serve to explain the principles of the invention. The drawings are only for the purpose of illustrating one or more preferred embodiments of the invention and are not to be construed as limiting the invention. In the drawings:

Fig. 1 is a block diagram of prior art conventional notch filtering of harmonic noise in a signal processing application;

Fig. 2 is a block diagram of the RANC signal processing of the present invention resulting in adaptive harmonic filtering;

Fig. 3 is a block diagram of the invention employed in a rate and/or acceleration control loop for adaptive structural resonance compensation;

Fig. 4 is a block diagram of the preferred RANC processing components and interface with the notch filter;

Fig. 5 is a block diagram of the preferred embodiment of the two pre-filters of the embodiment of Fig. 4;

Fig. 6 is a block diagram of the preferred demodulation, post-filter, and compensation elements of the embodiment of Fig. 4; and

Fig. 7 is a block diagram of a hardware-in-loop experimental implementation of the invention.

DETAILED DESCRIPTION OF THE INVENTION

The present invention is of an apparatus and method for control of mechanical structures with uncertain and/or time varying structural resonances (e.g., gimbaled turrets, seekers). The invention
5 employs notch filters to provide stability in the presence of uncertain and/or time varying resonances. Resonances are narrow in bandwidth, vary from system to system even with identical structure design, and may also vary with gimbal position, temperature, and/or vehicle maneuvers (high-g, g-vector). Accordingly, notch filters must also be narrow in bandwidth to minimize phase margin loss which implies that the resonance frequency must be known accurately.

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The present invention adaptively estimates the frequency of the resonance in realtime during normal closed-loop operation of the controller and compensates the notch filter to most effectively attenuate the resonance. The invention updates the frequency estimate to improve its accuracy from system to system -- in a given system, the invention corrects for errors in the initial estimate and adapts
15 to changes in frequency with time. Effectively, the invention also minimizes oscillations due to the resonance over the entire system operating range, which results in improved overall stabilization (e.g., for line-of-sight control for a gimbaled turret).

The present invention is preferably implemented in the controller as a stand-alone algorithm
20 running in parallel with a pre-existing conventional notch filter (Figs. 2-3). The RANC monitors both the input to the notch filter and its output. In return, the RANC compensates the notch filter with an updated estimate of the resonance frequency. The invention updates the estimate of ω_n by means of a pre-filtering, demodulation, and post-filtering means to generate a frequency error 70 (Fig. 4). Compensation is applied to the error signal to update the estimate of ω_n such that the notch output
25 error 80 (Fig. 5) is minimized. The RANC of the invention is preferably operated as a closed-loop system of, for example, about 10Hz bandwidth, and runs in parallel with the controller. The notch filter is thereby implemented as a time-varying filter with the natural frequency as an input variable instead of

a constant. The invention is fast enough to handle step changes in ω_n and yet remains robust in the presence of noise. The invention exploits the phase response of notch filters to locate ω_n . Near the notch frequency the phase is approximately linear and about the notch frequency the gain is approximately symmetrical.

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The preferred embodiment of the invention is now described with respect to Figs. 2-6. Fig. 2 shows the invention in comparison to the prior art shown in Fig. 1. Rather than assuming harmonic noise of constant frequency, the present invention accounts for a corrupted sensor **13** including
10 harmonic noise of time varying frequency $\omega_n(t)$. The RANC **22** and notch filter **14** employ an initial estimate **20** of the harmonic frequency, $\omega_n(0)$. The RANC samples both input and output of the notch filter and provides to the notch filter updated frequency estimate $\omega_n(k)$. Note that elements **20,22,14** may in fact be a plurality of such elements to handle a plurality of resonance features in a system.

15 A sample system employing RANC for structural resonance compensation is shown in Fig. 3, including rate command **30**, rate integrating gyroscope ("RIG") **32**, gyro noise **34**, compensation means **36**, drive command **38**, amplifier **40**, motor torque constant **42**, disturbance T_d **44**, gimbal with structural resonance **46**, integrator for gimbal rate (inertial) **48**, angular accelerometer **50**, accelerometer noise **52**, and feedback compensation **54**. An inherent characteristic of this application is that the
20 resonance amplitude trends to noise level as $\omega_n(k)$ approaches $\omega_n(t)$, thus rendering conventional techniques impractical.

Fig. 4 shows details of the preferred RANC of the invention in conjunction with notch input **72** and notch output **74**. The preferred RANC comprises first pre-filter **60**, second pre-filter **62**,
25 demodulator and post filter **64**, compensation means **66**, reference signal **68**, frequency update $\omega_n(k)$ **76**, and error signal e_ω **70**.

Fig. 5 shows details of the preferred pre-filters. The first pre-filter comprises zeta ratio means **84** and band pass filter **82** to provide notch error **80**. The zeta ratio attenuates the notch input **72** to match the attenuation of the notch filter at $\omega_n(k)$ such that the error at $\omega_n(k)$ approaches zero as $\omega_n(k)$ approaches the resonance frequency of the input signal. The second pre-filter comprises band pass filter **86** and low pass filter **88** to provide reference signal **68**. The band pass filters attenuate noise and disturbances at frequencies below and above $\omega_n(k)$ and the low pass filter produces a 90 degree phase lag at $\omega_n(k)$ between the reference **68** and notch error **80** by means of a second order filter.

Fig. 6 shows details of the preferred demodulator, post-filter, and compensation means. The demodulator and post-filter comprise multiplier **90**, low pass filter **92**, and sign detection function **94**. The compensation means comprises integral compensation **67**. The low pass filter is preferably second order with a bandwidth about one decade below the expected value of $\omega_n(k)$.

The advantages of the invention include the following: (1) The invention provides in a gimbaled turret or like structure a line-of-sight stabilization improvement over the entire range of gimbal position, temperature and vehicle maneuver. (2) The invention provides increased robustness to interchangeability of components and variations in assembly (e.g., preload, joint stiffness). (3) The invention relaxes stiffness and preload tolerances during assembly. (4) The invention helps reduce the size and weight of structures by relaxing the requirement that resonances be kept high in frequency. (5) The invention improves phase margin response –notches more narrow than conventional can be used. (6) The invention only requires a rough initial estimate of ω_n (in the example, within 35%, or $\pm 80\text{Hz}$). (7) The invention does not require an external reference signal. (8) The invention works independent of source type generating the resonance, i.e., the resonance source can be additive harmonic noise or mechanical structure. (9) The invention is computationally efficient for realtime applications (12-18th order overall). (10) The sampling period requirements of the invention are comparable to digital notch filters. (11) No spectral (FFT) or system identification techniques required. (12) The invention can be implemented with analog controllers and circuits. (13) The invention reduces the need for measuring

the structure of each particular system. (14) On systems that must be measured, the RANC simplifies the measurement process by automatically stabilizing the controller such that closed-loop measurements are possible. (15) The invention operates in a closed feedback loop ensuring that the notch cancels the resonance. (16) The invention is applicable, for example, to the compensation of
5 gyroscope spin frequency noise harmonics.

Features of the invention include the following: (1) The invention monitors notch error and updates the frequency estimate to most efficiently attenuate the resonance, in that for time-varying frequencies, the best estimate is not necessarily the instantaneous frequency. (2) The invention does
10 not require a constant input resonance amplitude, and so is useful for compensation of structural resonances where the main objective is to reduce the resonance to zero amplitude. (3) Multiple RANCs can be implemented in a single system, each operating in its own frequency range, and the designer can adjust the operating range in the pre-filters to minimize coupling between the RANCs. Reducing the frequency range of operation (by adjusting the pre-filters) allows for a larger number of
15 RANCs implemented simultaneously, each operating in its own frequency range. (4) The invention continues running in parallel with the controller to update the estimate of ω_n if the structural resonance changes for any reason, is not affected by controller input commands or uncorrelated external disturbances, and provides for straightforward implementation of graceful degradation logic. (5) The invention provides for a compact, modular design with minimal integration impact.

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The invention operates most efficiently under the following conditions: (1) Signal-to-Noise-Ratio ("S/N") > 1 for the resonance being identified (i.e., $S/N > 1$ near the frequency of interest, ω_n). If the S/N drops, the estimate remains relatively constant and is not updated until the S/N is again greater than 1. (2) No other uncompensated harmonics exist with $S/N > 1$ in the frequency range of operation (ω_n
25 $\pm 35\%$, or $\pm 80\text{Hz}$ in the example). Harmonics that do exist near ω_n can be filtered with conventional notch filters. Low frequency notches for such harmonics can be placed in the RANC closed-loop path

without affecting the rate or acceleration loops. Additional RANCs can be used if these harmonics are varying in frequency. (3) A sampling rate of at least 3 times ω_n is preferred for digital implementations.

Additional embodiments of the invention can be used for automatic notch depth adjustment in addition to frequency and for torque disturbance rejection of harmonic disturbances (e.g., disturbances caused by helicopter rotor frequency). Non-military applications of the present invention include 1) adaptive harmonic noise identification and filtering and 2) control of flexible structures. Specific potential applications are:

- 1) stabilization platforms;
- 2) gyroscopic rate sensors;
- 3) computer hard-drives;
- 4) vehicle body bending compensation; and
- 5) flexible manipulators and robotics.

Industrial Applicability:

The invention is further illustrated by the following non-limiting example.

Example 1

A "hardware-in-the-loop" system 100 as shown in Fig. 7 was implemented which provided for realtime simulation of a yaw/pitch and azimuth/elevation gimbal system. The RANC was implemented on yaw and pitch control loops with a 3600Hz update rate for the controller and gimbal model. Gimbals were modeled with one structural resonance each, yaw: 349Hz and pitch: 219Hz. The hardware was selected because it provides a significant source of electrical noise. The added components beyond those of Figs. 2-6 are gyro reference 102, hardware demodulation noise 104, analog-to-digital converters 106,118, digital-to-analog converters 108,116, amplifiers 110,114, summing node 120, and gyro demodulator 112.

With the RANC off, the system proved unstable with initial estimates of ω_n for yaw and pitch at 310Hz and 200Hz, respectively. With RANC on, the system was stable even with a step change from 325/200Hz (yaw/pitch) to 349/219Hz and could accommodate step rate commands, that is, the RANC operated properly despite the presence of step rate commands. Structural resonance amplitude was
5 reduced to noise level upon conversion of the RANC. For initial estimate errors within $\pm 20\%$, the invention showed a 0.2 sec settling time.

The preceding example can be repeated with similar success by substituting the generically or specifically described components and/or operating conditions of this invention for those used in the
10 preceding example.

Although the invention has been described in detail with particular reference to these preferred embodiments, other embodiments can achieve the same results. Variations and modifications of the present invention will be obvious to those skilled in the art and it is intended to cover in the appended
15 claims all such modifications and equivalents. The entire disclosures of all references, applications, patents, and publications cited above are hereby incorporated by reference.